

DENOISING OF 1-D SIGNAL USING DISCRETE WAVELET TRANSFORMS

Dr. Anil Kumar*

Associate Professor, Department of Physics, Hindu College, Moradabad-244001 (U.P.).

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*Corresponding Author

Dr. Anil Kumar

Associate Professor,
Department of Physics,
Hindu College, Moradabad-
244001 (U.P.).

ABSTRACT

Signal denoising is the process of removing noise from a signal for efficiency of getting information. The Signal denoising is implemented in software using MATLAB's Wavelet Toolbox. The study is carried out on 1-D bump signal. First of all we select the optimal base wavelet for the given signal. Thereafter Shannon Entropy cost function

thresholding is applied. This paper includes the discussion on the basics of wavelet, discrete wavelet transforms, selection of optimal wavelet, Shannon entropy cost function thresholding, signal denoising with results and conclusion.

KEYWORDS: Wavelet, Discrete Wavelet Transforms, Shannon Entropy, Signal Denoising.

INTRODUCTION

Wavelet is a new development in the emerging field of data analysis for Physicists, Engineers, and Environmentalists.^[1,2] It represents an efficient computational algorithm under the interest of a broad community. Fourier sine's extracts only frequency information from a time signal, thus losing time information.^[7] while wavelet extracts both time evolution and frequency composition of a signal. Wavelet is a special kind of the functions which exhibits oscillatory behaviour for a short time interval and then dies out. In wavelet we use a single function and its dilation and translation to generate a set of orthonormal basis functions to represent a signal. Number of such functions is infinite and we choose one that suits to our application. The range of interval over which scaling function and wavelet function are defined is known as support of wavelet. Beyond this interval (support) the functions should

be identically zero. There is an interesting relation between length of support and number of coefficients in the refinement relation. For orthogonal wavelet system, the length of support is always less than no. of coefficients in the refinement relation. It is also very helpful to require that the mother function have a certain number of zero moments, according to:

$$\int_{-\infty}^{\infty} \psi(t) dt = 0$$

The mother function can be used to generate a whole family of wavelets by translating and scaling the mother wavelet.

$$\psi_{a,b}(t) = \frac{1}{\sqrt{a}} \psi\left(\frac{t-b}{a}\right) = T_b D_a \psi$$

Here b is the translation parameter and a is the dilation or scaling parameter. Provided that $\psi(t)$ is real-valued, this collection of wavelets can be used as an orthonormal basis. A critical sampling of the continuous wavelet transform is

$$W_{a,b} = \int f(t) \frac{1}{\sqrt{a}} \psi\left(\frac{t-b}{a}\right) dt$$

is obtained via $a = 2^{-j}$, where j and k are integers representing the set of discrete translations and discrete dilations. Upon this substitution, we can write discrete wavelet transform as;

$$W_{j,k} = \int f(t) 2^{j/2} \psi(2^j t - k) dt$$

Wavelet coefficients for every (a, b) combination whereas in discrete wavelet transform, we find wavelet coefficients only at very few points by the dots and the wavelets that follow these values are given by:

$$\psi_{j,k}(t) = 2^{j/2} \psi(2^j t - k)$$

These wavelet coefficient for all j and k produce an orthonormal basis. We call $\psi_{0,0}(t) = \psi(t)$ as mother wavelet. Other wavelets are produced by translation and dilation of mother wavelet. The wavelet transform of a signal captures the localized time frequency information of the signal. Suppose we are given a signal or sequences of data $S = \{S_n\}_{n \in \mathbb{Z}}$ sampled at regular time interval $\Delta t = \tau$. S is split into a “blurred” version a_1 at the coarser interval $\Delta t = 2\tau$ and “detail” d_1 at scale $\Delta t = \tau$. This process is repeated and gives a sequence $S_n, a_1, a_2, a_3, a_4, \dots$ of more and more blurred versions together with the details $d_1, d_2, d_3, d_4, \dots$ removed at every scale ($\Delta t = 2^m \tau$ in a_m and d_m). Here a_m s and d_m s are

approximation and details of original signal. After N iteration the original signal S can be reconstructed as.

$$S = a_N + d_1 + d_2 + d_3 + \dots \dots d_N$$

Multiresolution Analysis and Discrete Wavelet Transforms

The discrete wavelet transform (DWT) provides a frequency band-wise decomposition of the signal, which is, called multiresolution analysis (MRA). A multiresolution analysis (MRA) for $L^2(\mathbb{R})$ introduced by Mallat.^[9,10] and extended by several researchers consists of a Sequence $V_j: j \in \mathbb{Z}$ of closed subspaces of $L^2(\mathbb{R})$ satisfying following properties;

$$(i) V_{j+1} \subset V_j: j \in \mathbb{Z}, \quad (2.1)$$

$$(ii) \bigcap_{j \in \mathbb{Z}} V_j = \{0\}, \quad \overline{\bigcup_{j \in \mathbb{Z}} V_j} = L^2(\mathbb{R}); \quad (2.2)$$

$$(iii) \text{ For every } L^2(\mathbb{R}), f(x) \in V_j \Rightarrow f\left(\frac{x}{2}\right) \in V_{j+1}, \quad \forall j \in \mathbb{Z} \quad \forall x \in \mathbb{R} \quad (2.3)$$

$$(iv) \text{ There exists a function } \varphi(x) \in V_0 \text{ such that } \{\varphi(x-k) : k \in \mathbb{Z}\} \text{ is orthonormal basis of } V_0. \quad (2.4)$$

The function whose existence is asserted in (3.7.4) is called a scaling function of the given MRA. The condition (3.7.4) is sometime relaxed by assuming that $\{\varphi(x-k) : k \in \mathbb{Z}\}$ is a Riesz basis for V_0 . That is, for every $f \in V_0$ there exists a unique sequences $\{h_k\}_{k \in \mathbb{Z}} \in \ell^2(\mathbb{Z})$ such that,

$$f(x) = \sum_{k \in \mathbb{Z}} h_k \varphi(x-k) \quad (2.5)$$

with convergence in $L^2(\mathbb{R})$.

To find an orthonormal wavelet, we need to do is to find a function $\psi \in W_0$ such that $\{\psi(x-k) : k \in \mathbb{Z}\}$ is an orthonormal basis of W_0 . In fact, if this is the case, then $\{2^{-j/2} \psi(2^{-j}x - k) : k \in \mathbb{Z}\}$ is an orthonormal basis for W_j for all $j \in \mathbb{Z}$. We can express function φ in terms of basis,

$$\varphi(x) = \frac{1}{\sqrt{2}} \sum_{k \in \mathbb{Z}} h_k \varphi\left(\frac{x}{2} - k\right) \quad (2.6)$$

where,

$$h_k = \sqrt{2} \int_{\mathbb{R}} \varphi(x) \overline{\varphi\left(\frac{x}{2} - k\right)} dx \quad \text{and} \quad \sum_{k \in \mathbb{Z}} |h_k|^2 < \infty.$$

The wavelet is a new analytical tool for turbulent or chaotic data to the physics community. It allows detection and characterization of short-lived structures in turbulence.

Optimal Base Wavelet and Denoising of Signal

The primary and most important work in the spectral analysis of any signal using wavelet transforms is the selection of suitable wavelet according to the signal.^[3,4] Suitable wavelet is selected on the basis of compatibility with signal characteristics. Accurate wavelet selection retains the original signal and also enhances the frequency spectrum of denoised signal.

Noise is the unwanted, problematic and unavoidable part of signal. A signal is represented as,

$$f = f_* + \sigma \cdot \rho \quad (3.1)$$

where f is the noise corrupted version of signal f_* and σ is the noise level and ρ is unit energy noise process. Here f_* is coherent and ρ is non-coherent with respect to optimal base wavelet.

Thresholding 3.2: A coherent signal is one that exhibits a concentration of energy in the representation domain and an incoherent signal is one whose energy is diffusely spread throughout the representation domain. A signal is coherent with respect to wavelet if the energy in the inner product representation is concentrated, that is, well localized in the representation domain. Thresholding is a technique performing to zero out small magnitude wavelet coefficients and retain the large magnitude wavelet.^[5,6] Signal noise ratio is the measurement of signal relative to noise and is described in terms of Shannon Entropy.

Shannon Entropy Cost function thresholding 3.3: Shannon Entropy is used to measure the amount of uncertainty in a probability distribution. Shannon Entropy Cost function is defined as,

$$M(\mathbf{c}, \{\mathbf{b}_j\}) = - \sum_{j=1}^M |\langle \mathbf{c}, \mathbf{b}_j \rangle|^2 \log |\langle \mathbf{c}, \mathbf{b}_j \rangle|^2$$

where $M = 2^N$, $1 \leq j \leq N$, $\mathbf{c} \in \mathbb{R}^M$. A best basis relative to M for \mathbf{c} is a system $\mathbf{B}^* \in B$ for which $M(\mathbf{c}, \mathbf{B}^*)$ is minimum.

Here

$$M(\mathbf{c}, \{\mathbf{b}_j\}) = \sum_{j: |\langle \mathbf{c}, \mathbf{b}_j \rangle| \geq 1} |\langle \mathbf{c}, \mathbf{b}_j \rangle|^2$$

Is a direct measure of mean square error encountered when the small (meaning below threshold) coefficients are discarded and the signal is reconstructed using the large (above threshold) coefficients.

The signal noise ratio (SNRAT) is measured in decibels and given by.

$$\text{SNRAT} = -10 \log_{10} (M(\mathbf{c} / \|\mathbf{c}\|, \{\mathbf{b}_j\})) \quad (3.4)$$

For a given threshold value $0 < \lambda$, we define,

$$M(\mathbf{c}, \{\mathbf{b}_j\}) = \left| \left\{ n : |\langle \mathbf{c}, \mathbf{b}_j \rangle| \geq \lambda \right\} \right|$$

In the context of signal processing cost function M measures how many coefficients are negligible (that is below threshold) in a transformed signal and how many are important.

The basis that concentrate the signal energy over a few coefficients, also reveals its time frequency structures, is called best basis. A best wavelet packet basis divides the time frequency plane into elementary atoms that are best adapted to approximate a particular signal. The best basis associated to a signal minimizes the Shannon Entropy function or Cost function M . Finding the minimum M , we require more than $2^{N/2}$ operations, which is computationally prohibitive. The fast dynamic programming algorithm of Coifman & Wickerhauser.^[8] find best basis with $O(N \log_2 N)$ operations by taking advantage of the tree structure.

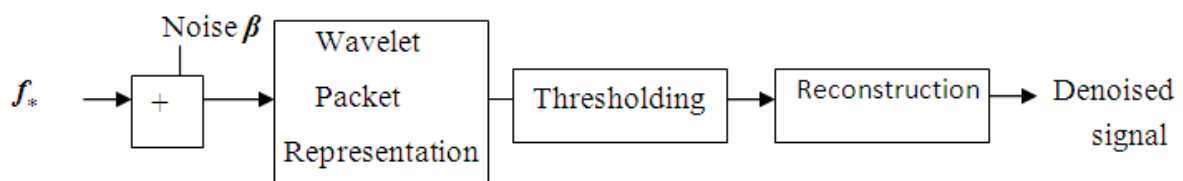


Figure 3.5 A block diagram for noise suppression and reconstruction algorithms.

RESULTS AND DISCUSSION

Let us consider a bumps signal for $N = 8$, so that the length of signal is $2^8 = 256$.

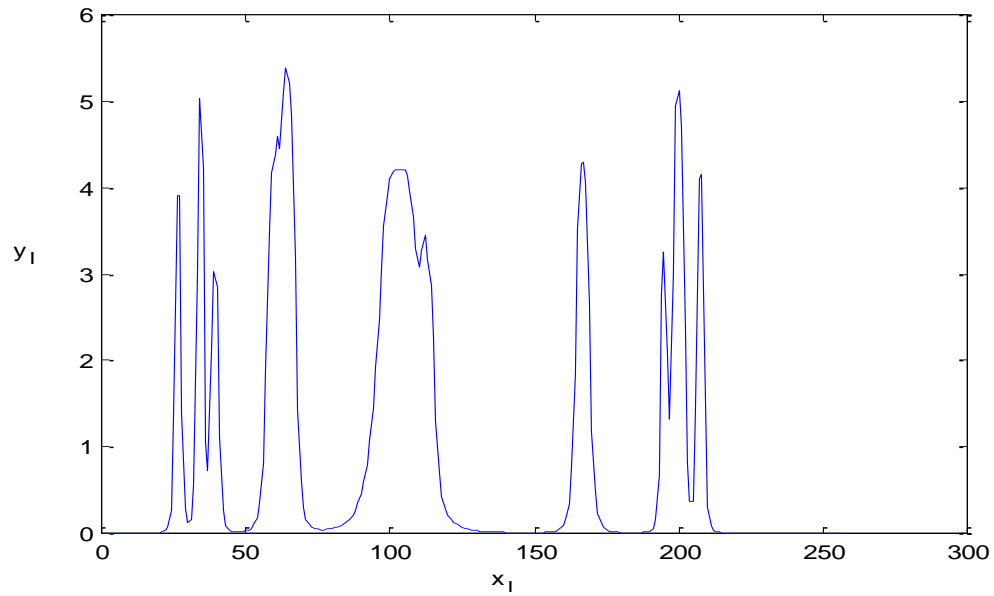


Figure 4.1: Bumps signal.

We add the noise to the above signal with signal noise ratio 1.5412.

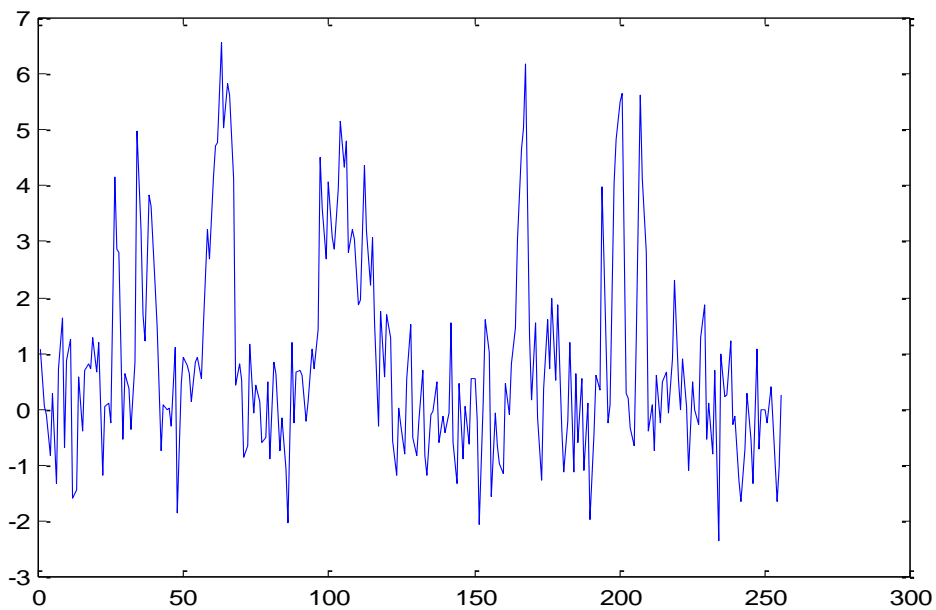


Figure 4.2: Noised signal.

We compute the discrete wavelet transform of the above noisy bumps signal using wavelet of Daubechies 4 wavelet, level

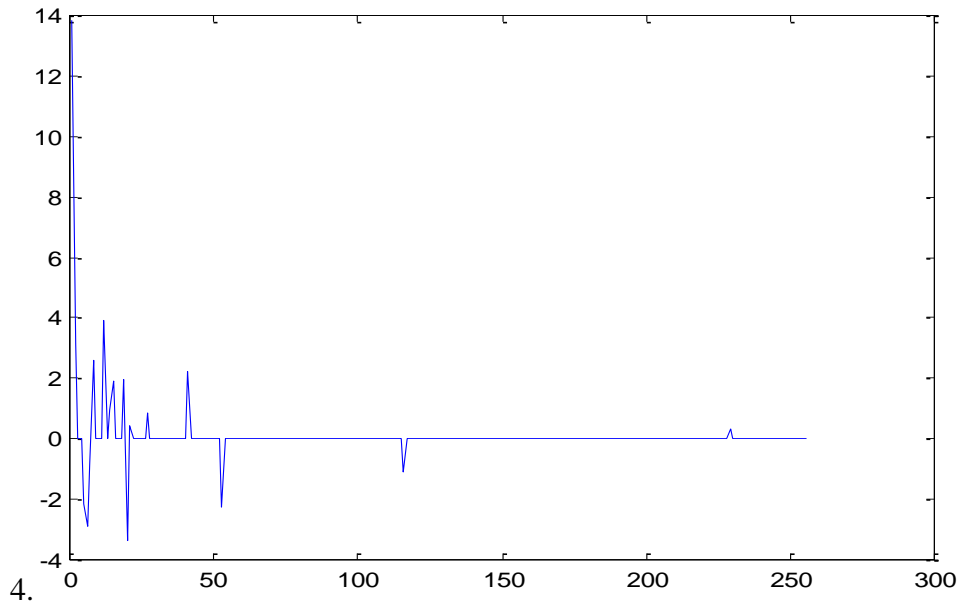


Figure 4.3: Wavelet coefficients for noised bumps signal.

Obviously, Most of the wavelet packet coefficients are nearly zero. Taking threshold value $\lambda = 0.7$ and using optimal base Daubechies 4 wavelet, level 4, we suppress noise and reconstruct the signal.

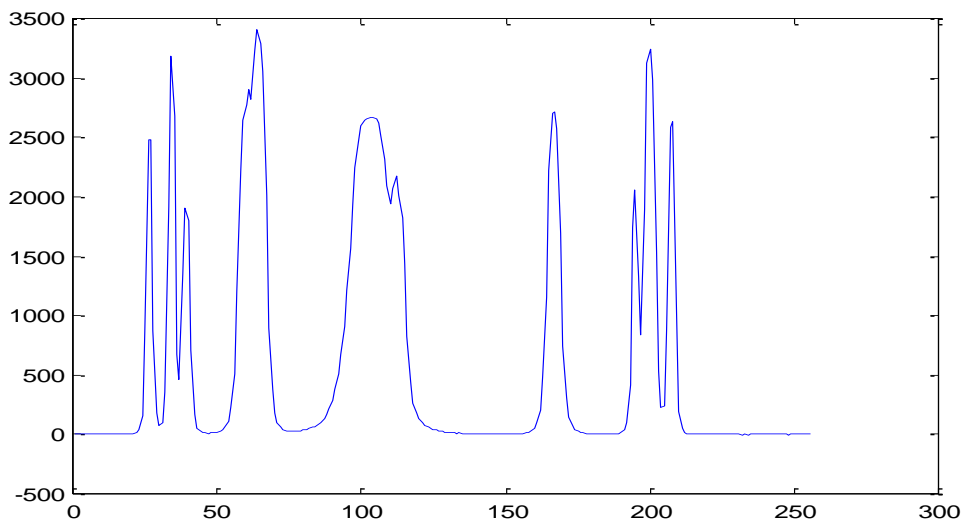


Figure 4.4: Denoised signal.

CONCLUSION

The discrete wavelet transforms provides a natural tool for denoising. Daubechies4 wavelet is the optimal base wavelet for given bump signal. Our approach of signal denoising is helpful for data compression as well as modulation and demodulation. Quantized coefficients below

threshold value are neglected and denoised signal is obtained as a version of input via an appropriate reconstruction algorithm.

BIBLIOGRAPHY

1. Kumar A, Kumar S and Pathak J K “Spectral Analysis of River Ramganga Hydraulics using Discrete wavelet transform”, International Conference of Advance Research and Innovation, 2015; 370-373.
2. Guerin Ch-A, Holchneider M Wavelet dimension and time evolution; Wavelets in Physics (Van Der Berg, J. C., ed.), 2004; 421-446.
3. Benedetto J J Irregular sampling and frames: Wavelets, A Tutorial in Theory and Applications, C. K. Chui, editor, Academic Press, Boston, 1992; 445-507.
4. Antoine J P Wavelet analysis: A new tool in Physics, Wavelets in Physics (Van Der Berg, J. C., ed.), 2004; 9-21.
5. Soman K P K and Ramchandran I Insight into wavelets: From theory to practice”, 2nd ed., Prentice Hall of India, 1998.
6. Pah N D and Kumar D K “Thresholding wavelet network for signal classification” International Journal of Wavelets, Multiresolution and Information Processing, 2003; 1(3): 243-261.
7. Duffin R J and Schaeffer A C “class of non-harmonic Fourier series”, Trans. Amer. Math. Soc., 1952; 72: 341-366.
8. Coifman R R, Meyer Y and Wickerhauser M V Wavelet analysis and signal processing; Wavelets and their Applications (M. B. Ruski et al., eds.), John and Bartlet Publishers, Boston, 1992; 153-178.
9. Mallat S G A theory for multiresolution signal decomposition: The wavelet representation”, IEEE Trans. On Pattern Analysis and Machine Intelligence, 1989; 11: 674-693.
10. Mallat S G “Multiresolution approximations and wavelet orthonormal bases of $L^2(\mathbb{R})$ ”, Trans. Amer. Math. Soc., 1989; 315: 69-87.